January 2000

LP2950/LP2951 Series of Adjustable Micropower Voltage Regulators

LP2950/LP2951 Series of Adjustable Micropower Voltage Regulators

General Description

The LP2950 and LP2951 are micropower voltage regulators with very low quiescent current (75 μ A typ.) and very low dropout voltage (typ. 40 mV at light loads and 380 mV at 100 mA). They are ideally suited for use in battery-powered systems. Furthermore, the quiescent current of the LP2950/LP2951 increases only slightly in dropout, prolonging battery life.

The LP2950-5.0 is available in the surface-mount D-Pak package, and in the popular 3-pin TO-92 package for pin-compatibility with older 5V regulators. The 8-lead LP2951 is available in plastic, ceramic dual-in-line, or metal can packages and offers additional system functions.

One such feature is an error flag output which warns of a low output voltage, often due to falling batteries on the input. It may be used for a power-on reset. A second feature is the logic-compatible shutdown input which enables the regulator to be switched on and off. Also, the part may be pin-strapped for a 5V, 3V, or 3.3V output (depending on the version), or programmed from 1.24V to 29V with an external pair of resistors.

Careful design of the LP2950/LP2951 has minimized all contributions to the error budget. This includes a tight initial tolerance (.5% typ.), extremely good load and line regulation (.05% typ.) and a very low output voltage temperature coefficient, making the part useful as a low-power voltage reference.

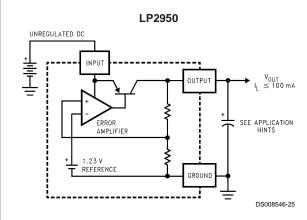
Features

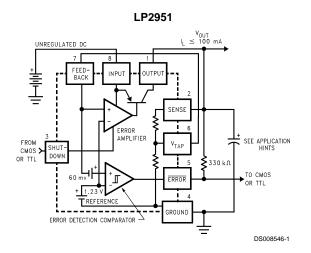
- 5V, 3V, and 3.3V versions available
- High accuracy output voltage
- Guaranteed 100 mA output current
- Extremely low quiescent current
- Low dropout voltage
- Extremely tight load and line regulation
- Very low temperature coefficient
- Use as Regulator or Reference
- Needs minimum capacitance for stability
- Current and Thermal Limiting
- Stable with low-ESR output capacitors

LP2951 versions only

- Error flag warns of output dropout
- Logic-controlled electronic shutdown
- Output programmable from 1.24 to 29V

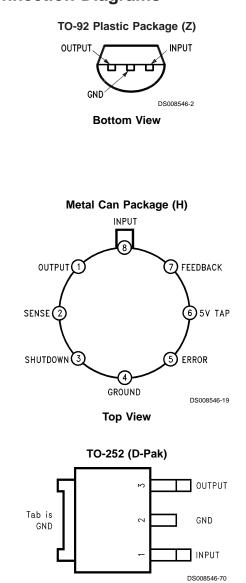
Block Diagram and Typical Applications



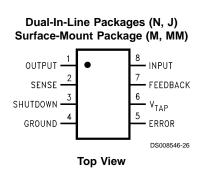




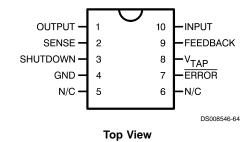
Connection Diagrams



Front View



10-Lead Ceramic Surface-Mount Package (WG)



www.national.com

Ordering Information

Package		Temperature					
	3.0V	3.3V	5.0V	(°C)			
TO-92 (Z)	LP2950ACZ-3.0	LP2950ACZ-3.3	LP2950ACZ-5.0	-40 < T _J < 125			
	LP2950CZ-3.0	LP2950CZ-3.3	LP2950CZ-5.0				
TO-252 (D-Pak)			LP2950CDT-5.0	$-40 < T_{J} < 125$			
N (N-08E)	LP2951ACN-3.0	LP2951ACN-3.3	LP2951ACN	-40 < T _J < 125			
	LP2951CN-3.0	LP2951CN-3.3	LP2951CN				
M (M08A)	LP2951ACM-3.0	LP2951ACM-3.3	LP2951ACM	-40 < T _J < 125			
	LP2951CM-3.0	LP2951CM-3.3	LP2951CM				
MM (MUA08A)	LP2951ACMM-3.0	LP2951ACMM-3.3	LP2951ACMM	-40 < T _J < 125			
	LP2951CMM-3.0	LP2951CMM-3.3	LP2951CMM				
J (J08A)			LP2951ACJ	$-40 < T_{J} < 125$			
			LP2951CJ				
			LP2951J	–55 < T _J < 150			
			LP2951J/883				
			5926-3870501MPA				
H (H08C)			LP2951H/883	–55 < T _J < 150			
			5962-3870501MGA				
WG (WG10A)			LP2951WG/883	–55 < T _J < 150			
			5962-3870501MXA				

Package Marking for MM Package:

Order Number	Package Marking
LP2951ACMM	LODA
LP2951CMM	LODB
LP2951ACMM-3.3	LOCA
LP2951CMM-3.3	LOCB
LP2951ACMM-3.0	LOBA
LP2951CMM-3.0	LOBB

Absolute Maximum Ratings (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

Input Supply Voltage	-0.3 to +30V
SHUTDOWN Input Voltage, Error Comparator Output Voltage, (Note 9)	
FEEDBACK Input Voltage	-1.5 to +30V
(Note 9) (Note 10)	
Power Dissipation	Internally Limited
Junction Temperature (T _J)	+150°C
Ambient Storage Temperature	–65° to +150°C

4 seconds, 260°C 10 seconds, 240°C 75 seconds, 219°C
TBD

Maximum Input Supply Voltage	30V
Junction Temperature Range (T _J) (Note 8)	
LP2951	–55° to +150°C
LP2950AC-XX, LP2950C-XX, LP2951AC-XX, LP2951C-XX	–40° to +125°C

Electrical Characteristics (Note 2)

		LP2951		LP2950AC-XX						
Parameter	Conditions (Note 2)				LP2951AC-XX			LP2951C-XX		
			Tested		Tested	Design		Tested	Design Limit	Units
		Тур	Limit	Тур	Limit	Limit	Тур	Limit		
			(Notes 3, 16)		(Note 3)	(Note 4)		(Note 3)	(Note 4)	
3V VERSIONS (Note 17)										
Output Voltage	$T_J = 25^{\circ}C$	3.0	3.015	3.0	3.015		3.0	3.030		V max
			2.985		2.985			2.970		V min
	$-25^{\circ}C \le T_{J} \le 85^{\circ}C$	3.0		3.0		3.030	3.0		3.045	V max
						2.970			2.955	V min
	Full Operating	3.0	3.036	3.0		3.036	3.0		3.060	V max
	Temperature Range		2.964			2.964			2.940	V min
Output Voltage	$100 \ \mu A \le I_L \le 100 \ mA$	3.0	3.045	3.0		3.042	3.0		3.072	V max
	$T_J \leq T_{JMAX}$		2.955			2.958			2.928	V min
3.3V VERSIONS (Note 17	7)									
Output Voltage	$T_J = 25^{\circ}C$	3.3	3.317	3.3	3.317		3.3	3.333		V max
			3.284		3.284			3.267		V mir
	$-25^{\circ}C \le T_{J} \le 85^{\circ}C$	3.3		3.3		3.333	3.3		3.350	V max
						3.267			3.251	V mir
	Full Operating	3.3	3.340	3.3		3.340	3.3		3.366	V max
	Temperature Range		3.260			3.260			3.234	V mir
Output Voltage	$100 \ \mu A \le I_L \le 100 \ mA$	3.3	3.350	3.3		3.346	3.3		3.379	V max
	$T_{J} \leq T_{JMAX}$		3.251			3.254			3.221	V mir
5V VERSIONS (Note 17)	•									
Output Voltage	$T_J = 25^{\circ}C$	5.0	5.025	5.0	5.025		5.0	5.05		V max
			4.975		4.975			4.95		V min
	$-25^{\circ}C \leq T_J \leq 85^{\circ}C$	5.0		5.0		5.05	5.0		5.075	V max
						4.95			4.925	V mir
	Full Operating	5.0	5.06	5.0		5.06	5.0		5.1	V max
	Temperature Range		4.94			4.94			4.9	V min
Output Voltage	$100 \ \mu A \le I_L \le 100 \ mA$	5.0	5.075	5.0		5.075	5.0		5.12	V max
	$T_{J} \leq T_{JMAX}$		4.925			4.925			4.88	V min
ALL VOLTAGE OPTION	S									
Output Voltage Temperature Coefficient	(Note 12)	20	120	20		100	50		150	ppm/°(
Line Regulation	$(V_0NOM + 1)V \le V_{in} \le 30V$ (Note 15)	0.03	0.1	0.03	0.1		0.04	0.2		% ma
(Note 14)			0.5			0.2			0.4	% ma
Load Regulation	$100 \ \mu A \le I_L \le 100 \ mA$	0.04	0.1	0.04	0.1		0.1	0.2		% ma
(Note 14)			0.3			0.2			0.3	% ma

			LP2951 LP2950AC-XX					LP2950C-	XX	
Parameter				LP2951AC-XX			LP2951C-XX			
	Conditions		Tested		Tested	Design		Tested	Design	Units
	(Note 2)	Тур	Limit	Тур	Limit	Limit	Тур	Limit	Limit	
			(Notes 3, 16)		(Note 3)	(Note 4)		(Note 3)	(Note 4)	
ALL VOLTAGE OPTION	S			1			1			I
Dropout Voltage (Note 5)	I _L = 100 μA		80		80			80		mV ma
(NOLE 5)		50	150	50		150	50		150	mV ma
	I _L = 100 mA		450		450			450		mV ma
		380	600	380		600	380		600	mV ma
Ground	I _L = 100 μA	75	120	75	120		75	120		µA ma
Current			140			140			140	µA ma
	I _L = 100 mA	8	12	8	12		8	12		mA ma
		_	14			14			14	mA ma
Dropout	$V_{in} = (V_O NOM - 0.5)V$	110	170	110	170		110	170		µA ma:
Ground Current	$I_{L} = 100 \ \mu A$		200	4.07		200	4.07	000	200	µA ma
Current Limit	$V_{out} = 0$	160	200	160	200		160	200		mA ma
	(Nata 40)	0.07	220	0.07		220	0.07		220	mA ma
Thermal Regulation	(Note 13)	0.05	0.2	0.05	0.2		0.05	0.2		%/W ma
Dutput Noise,	$C_L = 1 \ \mu F (5V \text{ Only})$	430		430			430			µV rm
10 Hz to 100 kHz	C _L = 200 µF	160		160			160			µV rm:
	$C_{L} = 3.3 \ \mu F$	100		100			100			
	(Bypass = $0.01 \mu F$	100		100			100			µV rm:
B-PIN VERSIONS ONLY	Pins 7 to 1 (LP2951))		LP2951		LP2951AC	• • • •		LP2951C-	vv	
Reference		1.235	1.25	1.235	1.25	-^^	1.235	1.26		V max
Voltage		1.200	1.26	1.200	1.20	1.26	1.200	1.20	1.27	V max
i onago			1.22		1.22			1.21		V min
			1.2			1.2		1.21	1.2	V min
Reference	(Note 7)		1.27			1.27			1.285	V max
Voltage			1.19			1.19			1.185	V min
Feedback Pin		20	40	20	40		20	40		nA max
Bias Current			60			60			60	nA max
Reference Voltage	(Note 12)	20		20			50			ppm/°C
Temperature Coefficient										
Feedback Pin Bias		0.1		0.1			0.1			nA/°C
Current Temperature										
Coefficient										
Error Comparator	L									1
Output Leakage	V _{OH} = 30V	0.01	1	0.01	1		0.01	1		µA ma:
Current			2			2			2	µA ma:
Output Low	$V_{in} = (V_O NOM - 0.5)V$	150	250	150	250		150	250		mV ma
Voltage	I _{OL} = 400 μA		400			400			400	mV ma
Upper Threshold	(Note 6)	60	40	60	40		60	40		mV mi
Voltage			25			25			25	mV mi
Lower Threshold	(Note 6)	75	95	75	95		75	95		mV ma
Voltage			140			140			140	mV ma
Hysteresis	(Note 6)	15		15			15			mV
Shutdown Input	Γ									
nput		1.3		1.3			1.3			V
Logic	Low (Regulator ON)		0.6			0.7			0.7	V max
Voltage	High (Regulator OFF)		2.0			2.0			2.0	V min

Electrical Characteristics (Note 2) (Continued)

Parameter	Conditions		LP2951		LP2950AC-XX LP2951AC-XX			LP2950C-XX LP2951C-XX		
	Conditions (Note 2)	Тур	Tested Limit (Notes 3, 16)	Тур	Tested Limit (Note 3)	Design Limit (Note 4)	Тур	Tested Limit (Note 3)	Design Limit (Note 4)	Units
Shutdown Input			x , , ,		. ,	()		()	()	
Shutdown Pin Input Current	V _{shutdown} = 2.4V	30	50	30	50		30	50		µA ma
			100			100			100	µA max
	V _{shutdown} = 30V	450	600	450	600		450	600		µA max
			750			750			750	µA max
Regulator Output Current in Shutdown	(Note 11)	3	10	3	10		3	10		µA max
			20			20			20	µA max

Note 1: Absolute Maximum Ratings are limits beyond which damage to the device may occur. Operating Ratings are conditions under which operation of the device is guaranteed. Operating Ratings do not imply guaranteed performance limits. For guaranteed performance limits and associated test conditions, see the Electrical Characteristics tables.

Note 2: Unless otherwise specified all limits guaranteed for $V_{IN} = (V_{ONOM} + 1)V$, $I_L = 100 \mu A$ and $C_L = 1 \mu F$ for 5V versions and 2.2 μF for 3V and 3.3V versions. Limits appearing in **boldface** type apply over the entire junction temperature range for operation. Limits appearing in normal type apply for $T_A = T_J = 25^{\circ}C$. Additional conditions for the 8-pin versions are FEEDBACK tied to V_{TAP} , OUTPUT tied to SENSE, and $V_{SHUTDOWN} \le 0.8V$.

Note 3: Guaranteed and 100% production tested.

Note 4: Guaranteed but not 100% production tested. These limits are not used to calculate outgoing AQL levels.

Note 5: Dropout Voltage is defined as the input to output differential at which the output voltage drops 100 mV below its nominal value measured at 1V differential. At very low values of programmed output voltage, the minimum input supply voltage of 2V (2.3V over temperature) must be taken into account.

Note 6: Comparator thresholds are expressed in terms of a voltage differential at the Feedback terminal below the nominal reference voltage measured at $V_{in} = (V_O NOM + 1)V$. To express these thresholds in terms of output voltage change, multiply by the error amplifier gain = $V_{out}/V_{ref} = (R1 + R2)/R2$. For example, at a programmed output voltage of 5V, the Error output is guaranteed to go low when the output drops by 95 mV x 5V/1.235V = 384 mV. Thresholds remain constant as a percent of V_{out} as V_{out} is varied, with the dropout warning occurring at typically 5% below nominal, 7.5% guaranteed.

Note 7: $V_{ref} \le V_{out} \le (V_{in} - 1V)$, 2.3V $\le V_{in} \le$ 30V, 100 $\mu A \le I_L \le$ 100 mA, $T_J \le T_{JMAX}$.

Note 8: The junction-to-ambient thermal resistances are as follows: 180°C/W and 160°C/W for the TO-92 package with 0.40 inch and 0.25 inch leads to the printed circuit board (PCB) respectively, 105°C/W for the molded plastic DIP (N), 130°C/W for the ceramic DIP (J), 160°C/W for the molded plastic SOP (M), 200°C/W for the molded plastic MSOP (MM), and 160°C/W for the metal can package (H). The above thermal resistances for the N, J, M, and MM packages apply when the package is soldered directly to the PCB. Junction-to-case thermal resistance for the H package is 20°C/W. Junction-to-case thermal resistance for the TO-252 package is 5.4°C/W.

Note 9: May exceed input supply voltage.

Note 10: When used in dual-supply systems where the output terminal sees loads returned to a negative supply, the output voltage should be diode-clamped to ground.

Note 11: $V_{shutdown} \ge 2V$, $V_{in} \le 30V$, V_{out} = 0, Feedback pin tied to V_{TAP} .

Note 12: Output or reference voltage temperature coefficient is defined as the worst case voltage change divided by the total temperature range.

Note 13: Thermal regulation is defined as the change in output voltage at a time T after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a 50 mA load pulse at V_{IN} = 30V (1.25W pulse) for T = 10 ms.

Note 14: Regulation is measured at constant junction temperature, using pulse testing with a low duty cycle. Changes in output voltage due to heating effects are covered under the specification for thermal regulation.

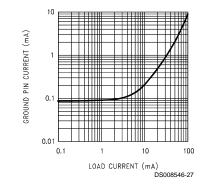
Note 15: Line regulation for the LP2951 is tested at 150°C for $I_L = 1$ mA. For $I_L = 100 \ \mu$ A and $T_J = 125°C$, line regulation is guaranteed by design to 0.2%. See Typical Performance Characteristics for line regulation versus temperature and load current.

Note 16: A Military RETS specification is available on request. At time of printing, the LP2951 RETS specification complied with the boldface limits in this column. The LP2951H, WG, or J may also be procured as Standard Military Drawing Spec #5962-3870501MGA, MXA, or MPA.

Note 17: All LP2950 devices have the nominal output voltage coded as the last two digits of the part number. In the LP2951 products, the 3.0V and 3.3V versions are designated by the last two digits, but the 5V version is denoted with no code at this location of the part number (refer to ordering information table).

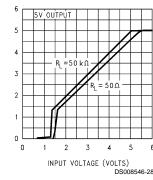
Typical Performance Characteristics

Quiescent Current



Dropout Characteristics

OUTPUT VOLTAGE (VOLTS)

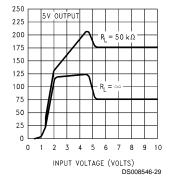


Input Current

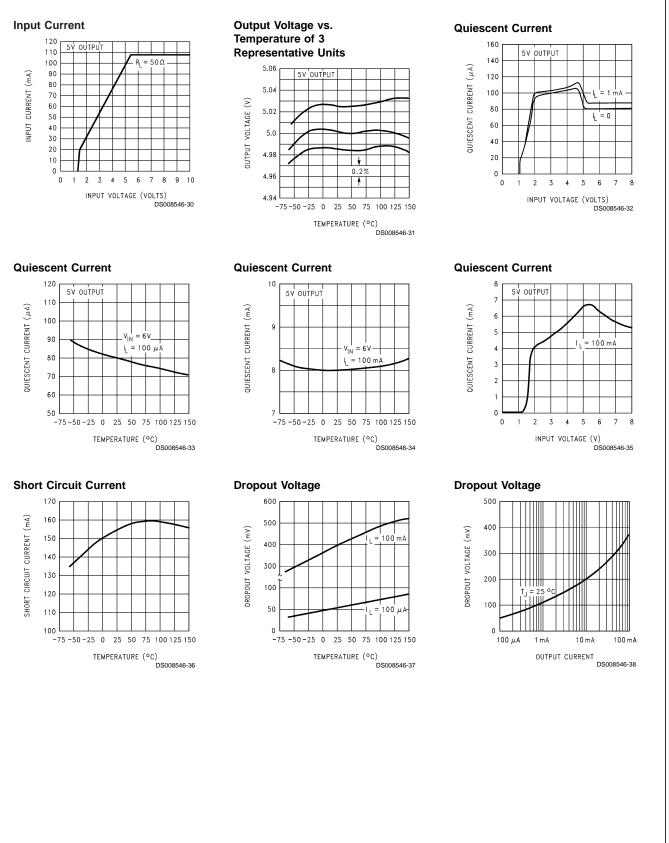
(μμ)

CURRENT

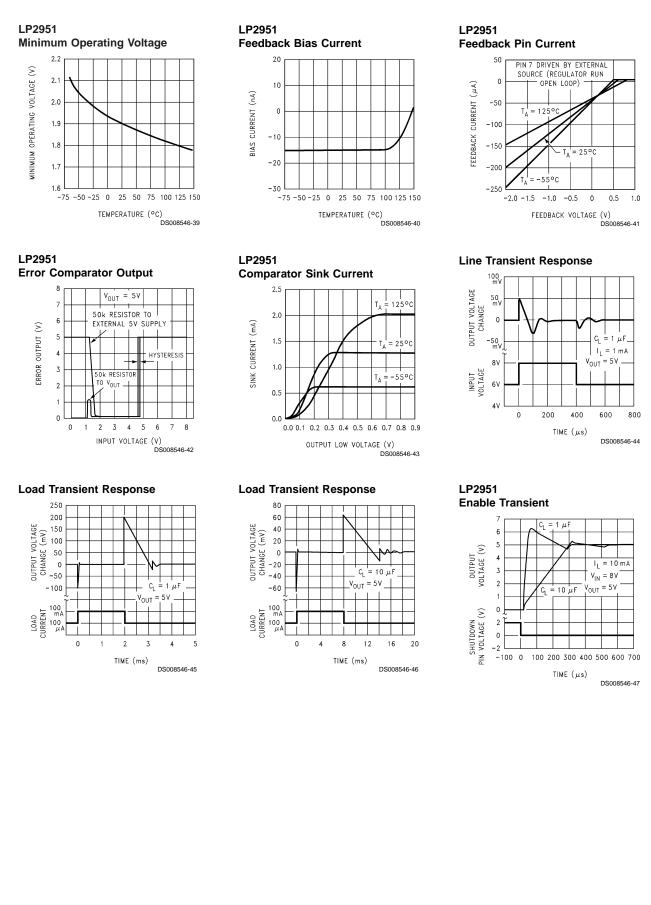
INPUT



Typical Performance Characteristics (Continued)

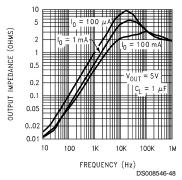


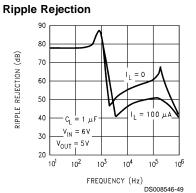
Typical Performance Characteristics (Continued)



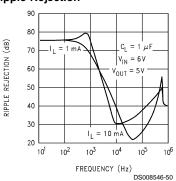
Typical Performance Characteristics (Continued)

Output Impedance

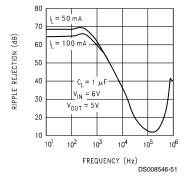




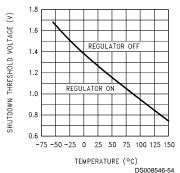




Ripple Rejection



Shutdown Threshold Voltage



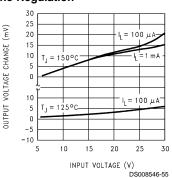
VOLTAGE NOISE SPECTRAL DENSITY ($\mu v M H_z$ 3.0 2.5 = 220 2.0 1.5 1.0 0.5 0.0 10² 10³ 10 105 FREQUENCY (Hz)DS008546-52

=100 mA

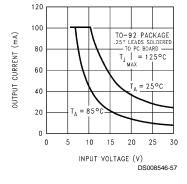


LP2951 Output Noise

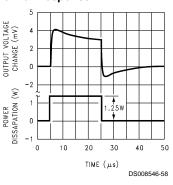
3.5



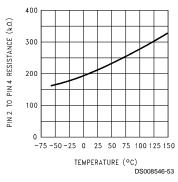
LP2950 Maximum Rated Output Current



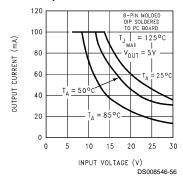
Thermal Response



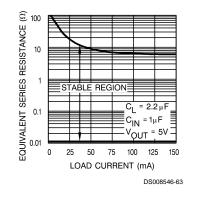
LP2951 Divider Resistance



LP2951 Maximum Rated Output Current



Output Capacitor ESR Range



Application Hints

EXTERNAL CAPACITORS

A 1.0 μ F (or greater) capacitor is required between the output and ground for stability at output voltages of 5V or more. At lower output voltages, more capacitance is required (2.2 μ F or more is recommended for 3V and 3.3V versions). Without this capacitor the part will oscillate. Most types of tantalum or aluminum electrolytics work fine here; even film types work but are not recommended for reasons of cost. Many aluminum electrolytics have electrolytes that freeze at about -30°C, so solid tantalums are recommended for operation below -25°C. The important parameters of the capacitor are an ESR of about 5 Ω or less and a resonant frequency above 500 kHz. The value of this capacitor may be increased without limit.

At lower values of output current, less output capacitance is required for stability. The capacitor can be reduced to 0.33 μ F for currents below 10 mA or 0.1 μ F for currents below 1 mA. Using the adjustable versions at voltages below 5V runs the error amplifier at lower gains so that *more* output capacitance is needed. For the worst-case situation of a 100 mA load at 1.23V output (Output shorted to Feedback) a 3.3 μ F (or greater) capacitor should be used.

Unlike many other regulators, the LP2950 will remain stable and in regulation with no load in addition to the internal voltage divider. This is especially important in CMOS RAM keep-alive applications. When setting the output voltage of the LP2951 versions with external resistors, a minimum load of 1 μ A is recommended.

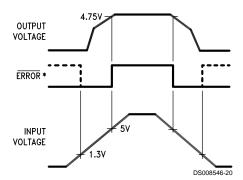
A 1 μ F tantalum or aluminum electrolytic capacitor should be placed from the LP2950/LP2951 input to ground if there is more than 10 inches of wire between the input and the AC filter capacitor or if a battery is used as the input.

Stray capacitance to the LP2951 Feedback terminal can cause instability. This may especially be a problem when using high value external resistors to set the output voltage. Adding a 100 pF capacitor between Output and Feedback and increasing the output capacitor to at least 3.3 μ F will fix this problem.

ERROR DETECTION COMPARATOR OUTPUT

The comparator produces a logic low output whenever the LP2951 output falls out of regulation by more than approximately 5%. This figure is the comparator's built-in offset of about 60 mV divided by the 1.235 reference voltage. (Refer to the block diagram in the front of the datasheet.) This trip level remains "5% below normal" regardless of the programmed output voltage of the 2951. For example, the error flag trip level is typically 4.75V for a 5V output or 11.4V for a 12V output. The out of regulation condition may be due either to low input voltage, current limiting, or thermal limiting. Figure 1 below gives a timing diagram depicting the ERROR signal and the regulated output voltage as the LP2951 input is ramped up and down. For 5V versions, the ERROR signal becomes valid (low) at about 1.3V input. It goes high at about 5V input (the input voltage at which V_{OUT} = 4.75V). Since the LP2951's dropout voltage is load-dependent (see curve in typical performance characteristics), the input voltage trip point (about 5V) will vary with the load current. The output voltage trip point (approx. 4.75V) does not vary with load.

The error comparator has an open-collector output which requires an external pullup resistor. This resistor may be returned to the output or some other supply voltage depending on system requirements. In determining a value for this resistor, note that while the output is rated to sink 400 $\mu A,$ this sink current adds to battery drain in a low battery condition. Suggested values range from 100k to 1 M $\Omega.$ The resistor is not required if this output is unused.



*When $V_{IN} \leq 1.3V$, the error flag pin becomes a high impedance, and the error flag voltage rises to its pull-up voltage. Using V_{OUT} as the pull-up voltage (see *Figure 2*), rather than an external 5V source, will keep the error flag voltage under 1.2V (typ.) in this condition. The user may wish to divide down the error flag voltage using equal-value resistors (10 k Ω suggested), to ensure a low-level logic signal during any fault condition, while still allowing a valid high logic level during normal operation.

FIGURE 1. ERROR Output Timing

PROGRAMMING THE OUTPUT VOLTAGE (LP2951)

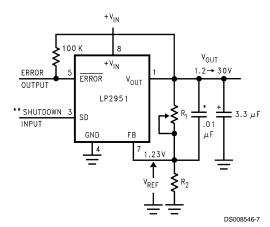
The LP2951 may be pin-strapped for the nominal fixed output voltage using its internal voltage divider by tying the output and sense pins together, and also tying the feedback and V_{TAP} pins together. Alternatively, it may be programmed for any output voltage between its 1.235V reference and its 30V maximum rating. As seen in *Figure 2*, an external pair of resistors is required.

The complete equation for the output voltage is

$$V_{OUT} = V_{REF} \bullet \left(1 + \frac{R_1}{R_2}\right) + I_{FB}R_1$$

where V_{REF} is the nominal 1.235 reference voltage and I_{FB} is the feedback pin bias current, nominally –20 nA. The minimum recommended load current of 1 µA forces an upper limit of 1.2 MΩ on the value of R_2 , if the regulator must work with no load (a condition often found in CMOS in standby). I_{FB} will produce a 2% typical error in V_{OUT} which may be eliminated at room temperature by trimming R_1 . For better accuracy, choosing R_2 = 100k reduces this error to 0.17% while increasing the resistor program current to 12 µA. Since the LP2951 typically draws 60 µA at no load with Pin 2 open-circuited, this is a small price to pay.

Application Hints (Continued)



*See Application Hints

 $V_{out} = V_{Ref} \left(1 + \frac{R_1}{R_2} \right)$

Drive with TTL-high to shut down. Ground or leave open if shutdown feature is not to be used. **Note: Pins 2 and 6 are left open.

FIGURE 2. Adjustable Regulator

REDUCING OUTPUT NOISE

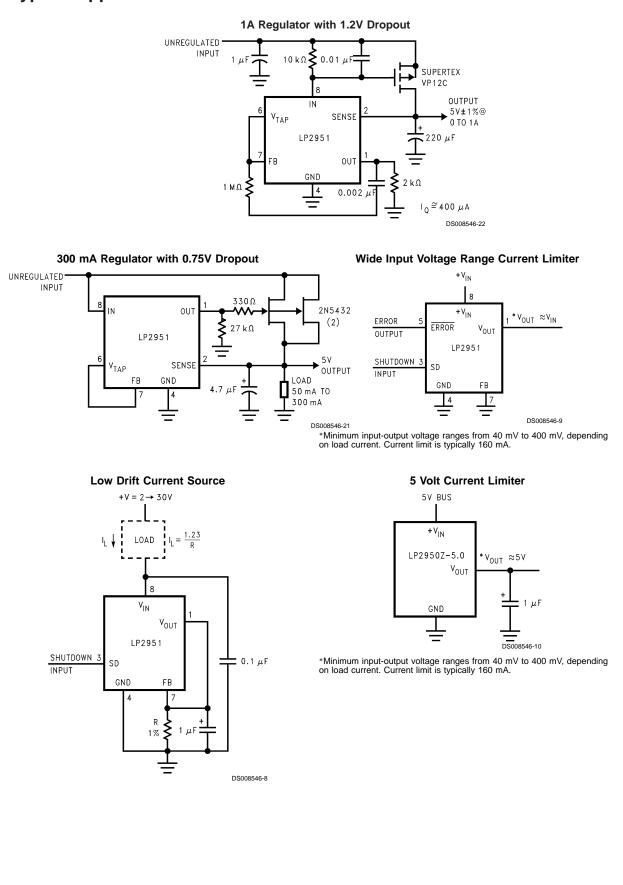
In reference applications it may be advantageous to reduce the AC noise present at the output. One method is to reduce the regulator bandwidth by increasing the size of the output capacitor. This is the only way noise can be reduced on the 3 lead LP2950 but is relatively inefficient, as increasing the capacitor from 1 μF to 220 μF only decreases the noise from 430 μV to 160 μV rms for a 100 kHz bandwidth at 5V output.

Noise can be reduced fourfold by a bypass capacitor across $\mathsf{R}_1,$ since it reduces the high frequency gain from 4 to unity. Pick

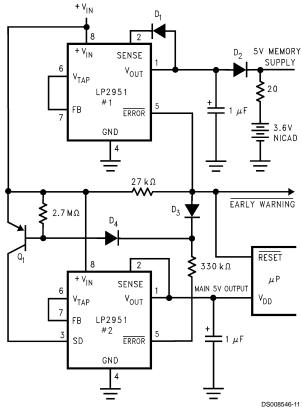
$$C_{BYPASS} \cong \frac{1}{2\pi R_1 \bullet 200 \text{ Hz}}$$

or about 0.01 μ F. When doing this, the output capacitor must be increased to 3.3 μ F to maintain stability. These changes reduce the output noise from 430 μ V to 100 μ V rms for a 100 kHz bandwidth at 5V output. With the bypass capacitor added, noise no longer scales with output voltage so that improvements are more dramatic at higher output voltages.

Typical Applications



Regulator with Early Warning and Auxiliary Output



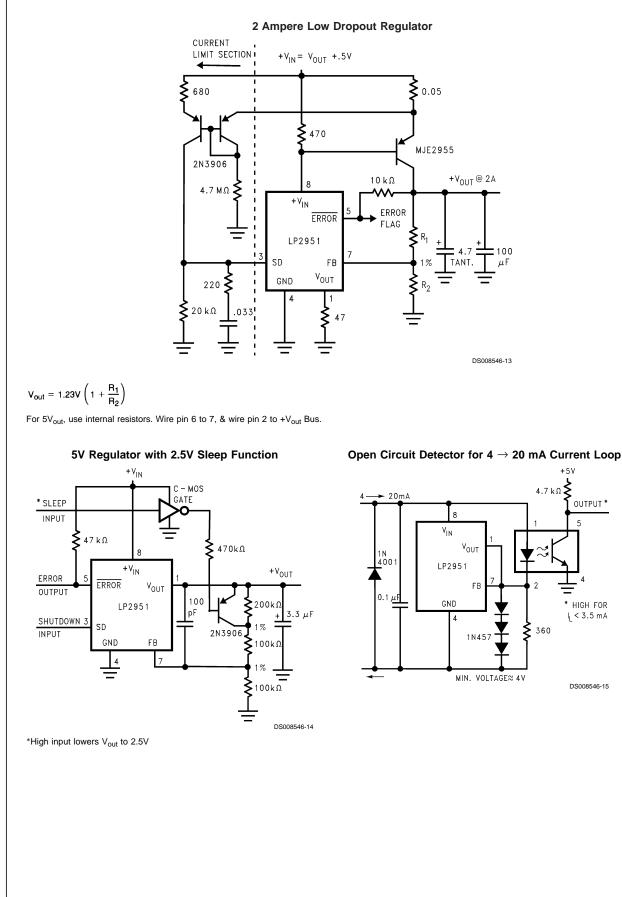
Latch Off When Error Flag Occurs +V_{in} 8 ≶ 470 kΩ + V_{IN} V_{OUT} ERROR Ś V_{OUT} 470 kΩ ≶ LP2951 R₁ 1μF SD FΒ D C RESET GND 3 R₂ 4 DS008546-12 LP2950/LP2951

3000340-11

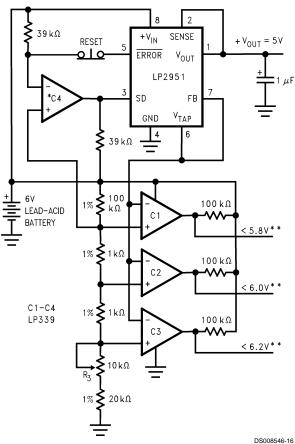
Early warning flag on low input voltageMain output latches off at lower input voltages

Battery backup on auxiliary output

Detection: Reg. #1's V_{out} is programmed one diode drop above 5V. Its error flag becomes active when V_{in} \leq 5.7V. When V_{in} drops below 5.3V, the error flag of Reg. #2 becomes active and via Q1 latches the main output off. When V_{in} again exceeds 5.7V Reg. #1 is back in regulation and the early warning signal rises, unlatching Reg. #2 via D3.

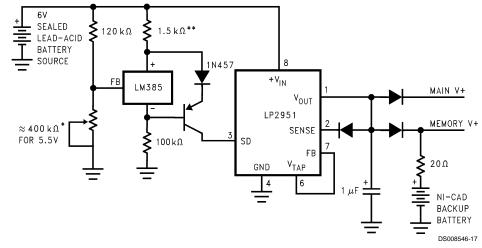


Regulator with State-of-Charge Indicator



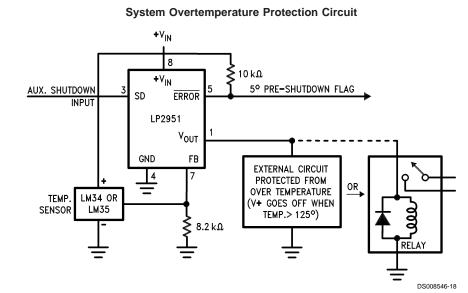
*Optional Latch off when drop out occurs. Adjust R3 for C2 Switching when V_{in} is 6.0V. **Outputs go low when V_{in} drops below designated thresholds.

Low Battery Disconnect

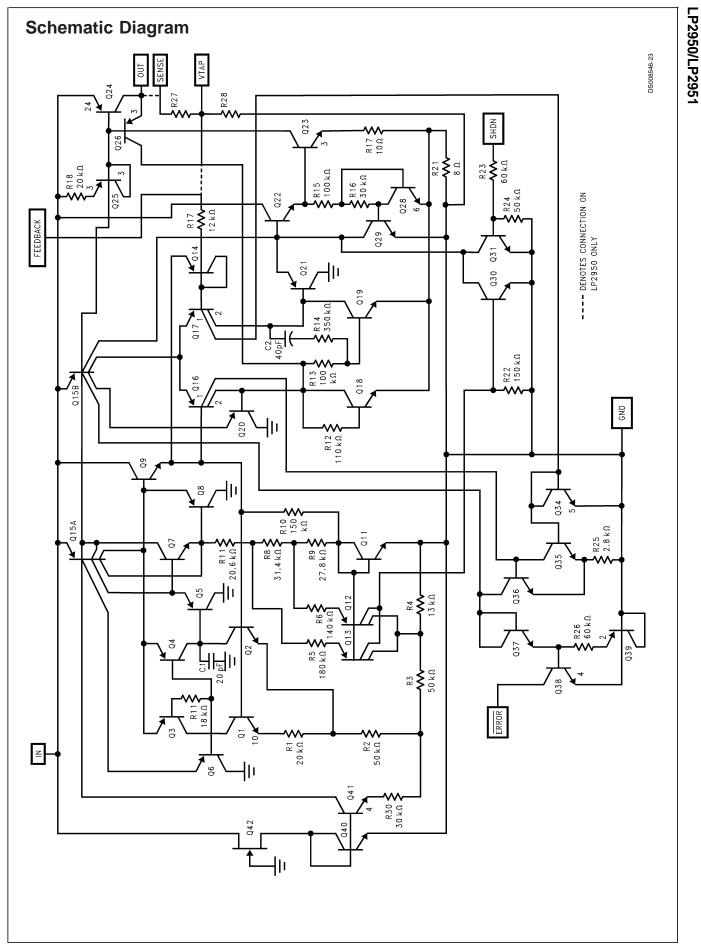


For values shown, Regulator shuts down when $V_{in} < 5.5V$ and turns on again at 6.0V. Current drain in disconnected mode is $\approx 150 \mu$ A. *Sets disconnect Voltage

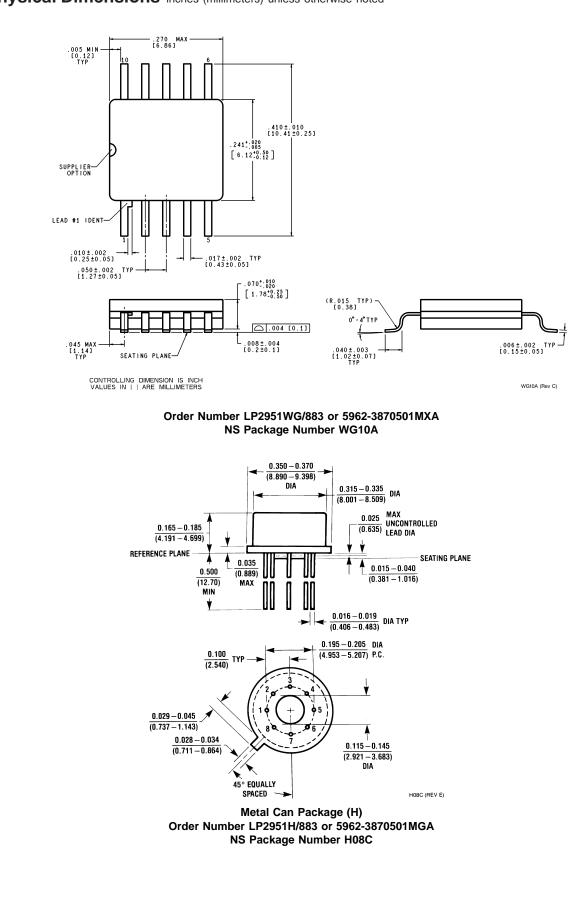
**Sets disconnect Hysteresis



LM34 for 125°F Shutdown LM35 for 125°C Shutdown

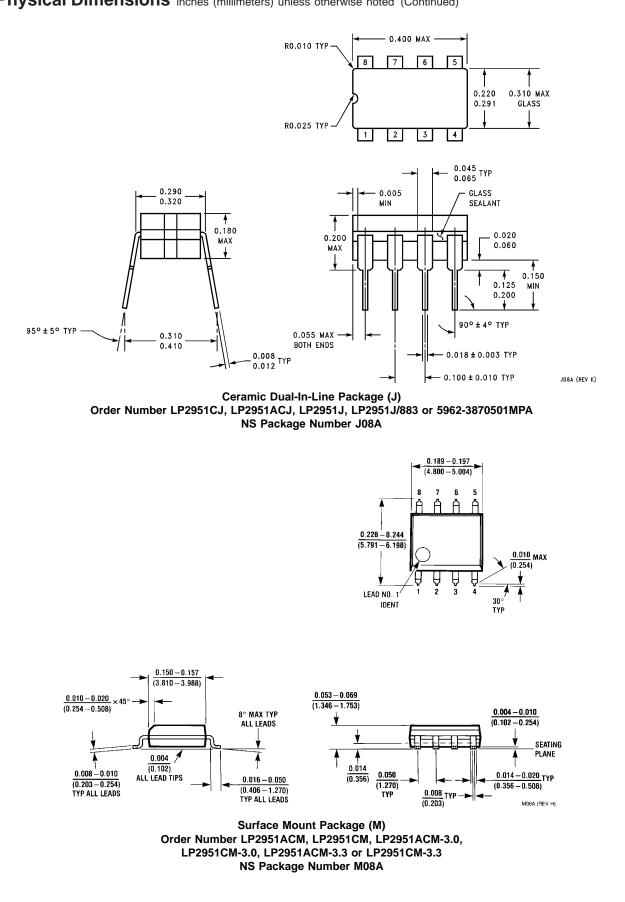


Physical Dimensions inches (millimeters) unless otherwise noted



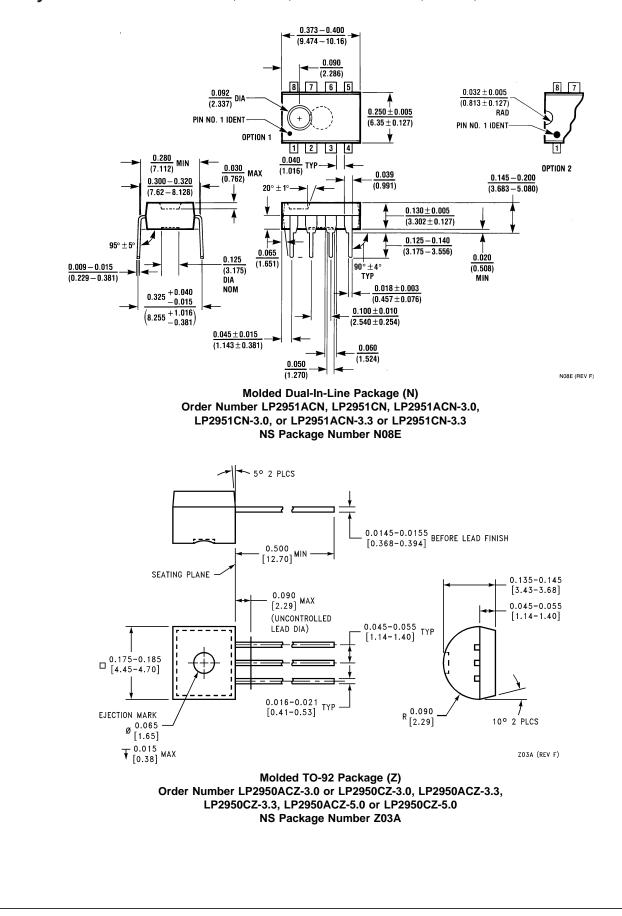


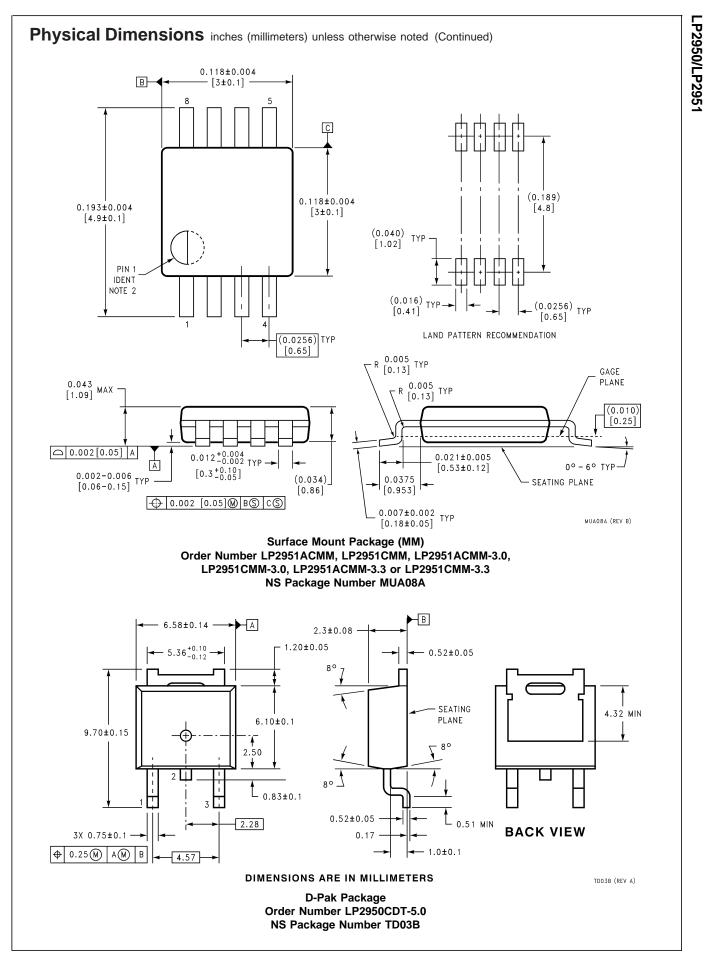
Physical Dimensions inches (millimeters) unless otherwise noted (Continued)





Physical Dimensions inches (millimeters) unless otherwise noted (Continued)





Notes

LIFE SUPPORT POLICY

NATIONAL'S PRODUCTS ARE NOT AUTHORIZED FOR USE AS CRITICAL COMPONENTS IN LIFE SUPPORT DEVICES OR SYSTEMS WITHOUT THE EXPRESS WRITTEN APPROVAL OF THE PRESIDENT AND GENERAL COUNSEL OF NATIONAL SEMICONDUCTOR CORPORATION. As used herein:

- Life support devices or systems are devices or systems which, (a) are intended for surgical implant into the body, or (b) support or sustain life, and whose failure to perform when properly used in accordance with instructions for use provided in the labeling, can be reasonably expected to result in a significant injury to the user.
- A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.

 National Semiconductor Corporation Americas Tel: 1-800-272-9959
 National Semiconductor Europe
 National Semiconductor
 National Semiconductor

 Fax: 1-800-737-7018
 Fax: +49 (0) 1 80-530 85 86
 Res

 Email: support@nsc.com
 Tel: +49 (0) 1 80-530 85 85
 Fax

 Email: support@nsc.com
 English
 Tel: +49 (0) 1 80-530 85 85
 Fax

 www.national.com
 Italiano
 Tel: +49 (0) 1 80-532 93 58
 Email

National Semiconductor Asia Pacific Customer Response Group Tel: 65-2544466 Fax: 65-2504466 Email: sea.support@nsc.com National Semiconductor Japan Ltd. Tel: 81-3-5639-7560 Fax: 81-3-5639-7507

National does not assume any responsibility for use of any circuitry described, no circuit patent licenses are implied and National reserves the right at any time without notice to change said circuitry and specifications.