

UCSP, Single-Supply, Low-Noise, Low-Distortion, Rail-to-Rail Op Amps

General Description

The MAX4249–MAX4257 low-noise, low-distortion operational amplifiers offer rail-to-rail outputs and single-supply operation down to 2.4V. They draw 400 μ A of quiescent supply current per amplifier while featuring ultra-low distortion (0.0002% THD), as well as low input voltage-noise density (7.9nV/ $\sqrt{\text{Hz}}$) and low input current-noise density (0.5fA/ $\sqrt{\text{Hz}}$). These features make the devices an ideal choice for portable/battery-powered applications that require low distortion and/or low noise.

For additional power conservation, the MAX4249/ MAX4251/MAX4253/MAX4256 offer a low-power shutdown mode that reduces supply current to 0.5µA and puts the amplifiers' outputs into a high-impedance state. The MAX4249-MAX4257's outputs swing rail-torail and their input common-mode voltage range includes ground. The MAX4250-MAX4254 are unitygain stable with a gain-bandwidth product of 3MHz. The MAX4249/MAX4255/MAX4256/MAX4257 are internally compensated for gains of 10V/V or greater with a gain-bandwidth product of 22MHz. The single MAX4250/ MAX4255 are available in space-saving 5-pin SOT23 packages. The MAX4252 is available in an 8-bump chipscale package (UCSPTM) and the MAX4253 is available in a 10-bump UCSP. The MAX4250AAUK comes in a 5-pin SOT23 package and is specified for operation over the automotive (-40°C to +125°C) temperature range.

Applications

Wireless Communications Devices

PA Control

Portable/Battery-Powered Equipment

Medical Instrumentation

ADC Buffers

Digital Scales/Strain Gauges

Features

- ◆ Available in Space-Saving UCSP, SOT23, and µMAX[®] Packages
- ♦ Low Distortion: 0.0002% THD (1kΩ load)
- ♦ 400µA Quiescent Supply Current per Amplifier
- ♦ Single-Supply Operation from 2.4V to 5.5V
- ♦ Input Common-Mode Voltage Range Includes Ground
- ♦ Outputs Swing Within 8mV of Rails with a 10kΩ Load
- ◆ 3MHz GBW Product, Unity-Gain Stable (MAX4250–MAX4254)

22MHz GBW Product, Stable with Ay ≥ 10V/V (MAX4249/MAX4255/MAX4256/MAX4257)

- ◆ Excellent DC Characteristics V_{OS} = 70μV I_{BIAS} = 1pA Large-Signal Voltage Gain = 116dB
- Low-Power Shutdown Mode Reduces Supply Current to 0.5μA Places Outputs in a High-Impedance State
- ♦ 400pF Capacitive-Load Handling Capability

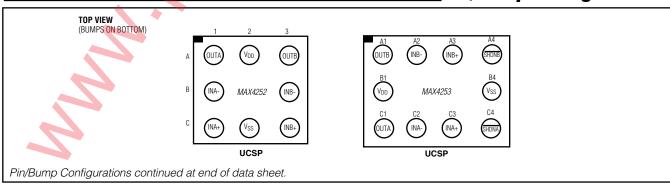
Ordering Information

PART	TEMP RANGE	PIN- PACKAGE	TOP MARK
MAX4249ESD+	-40°C to +85°C	14 SO	
MAX4249EUB+	-40°C to +85°C	10 μMAX	_
MAX4250EUK+T	-40°C to +85°C	5 SOT23	ACCI
MAX4250AAUK+T	-40°C to +125°C	5 SOT23	AEYJ

⁺Denotes a lead(Pb)-free/RoHS-compliant package. T = Tape and reel.

Ordering Information continued at end of data sheet. Selector Guide appears at end of data sheet.

Pin/Bump Configurations



UCSP is a trademark and µMAX is a registered trademark of Maxim Integrated Products, Inc.

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ABSOLUTE MAXIMUM RATINGS

Power-Supply Voltage (V _{DD} to V _{SS})+6.0V to -0.3V
Analog Input Voltage (IN_+, IN)(V _{DD} + 0.3V) to (V _{SS} - 0.3V)
SHDN Input Voltage
Output Short-Circuit Duration to Either SupplyContinuous
Continuous Power Dissipation (T _A = +70°C)
5-Pin SOT23 (derate 7.1mW/°C above +70°C)571mW
8-Bump UCSP (derate 4.7mW/°C above +70°C)379mW
8-Pin µMAX (derate 4.5mW/°C above +70°C)362mW
8-Pin SO (derate 5.88mW/°C above +70°C)471mW
10-Bump UCSP (derate 6.1mW/°C above +70°C)484mW

10-Pin µMAX (derate 5.6mW/°C above 14-Pin SO (derate 8.33mW/°C above + Operating Temperature Range	667mW 40°C to +85°C 40°C to +125°C +150°C 65°C to +150°C +300°C
Soldering Temperature (reflow)	

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

 $(V_{DD} = 5V, V_{SS} = 0V, V_{CM} = 0V, V_{OUT} = V_{DD}/2, R_L \text{ connected to } V_{DD}/2, \overline{SHDN} = V_{DD}, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted.}$ Typical values are at $T_A = +25^{\circ}C$.) (Notes 2, 3)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS		
Supply Voltage Range	V _{DD}	(Note 4)	(Note 4)		2.4		5.5	V	
			$V_{DD} = 3V$			400			
		Normal	\/	E temperature		420	575		
Quiescent Supply Current Per Amplifier	IQ	mode	$V_{DD} = 5V$	MAX4250AAUK			675	μΑ	
Amplillel			V _{DD} = 5V, U	CSP only		420	655		
		Shutdow	n mode (SHD	N = V _{SS}) (Note 2)		0.5	1.5		
	.,,	E temperature			±0.07	±0.75			
Input Offset Voltage (Note 5)	Vos	MAX425	MAX4250AAUK				±1.85	mV	
Input Offset Voltage Tempco	TCV _{OS}					0.3		μV/°C	
	IB		$T_A = +25$	T _A = +25°C		0.1	1		
Input Bias Current		(Note 6)	T _A = -40°C to +85°C				50	рА	
			$T_A = -40^{\circ}C \text{ to } +125^{\circ}C$				1500		
			$T_A = +25^{\circ}C$			0.1	1		
Input Offset Current	los	(Note 6)	$T_A = -40^{\circ}$	°C to +85°C			10	рА	
			$T_A = -40^{\circ}$	°C to +125°C			100		
Differential Input Resistance	R _{IN}				1000		$G\Omega$		
Input Common-Mode Voltage	Voi	Guaranteed by E temperature		-0.2		V _{DD} -1.1	V		
Range	V _{CM}	CMRR test MAX4250AAUK		0		V _{DD} -1.1	V		
Common-Mode Rejection Ratio	CMRR	V _{SS} - 0.2	$2V \le V_{CM} \le$	E temperature	70	115		dB	
Common-wode nejection Ratio	Civinn	V _{DD} - 1.1V		MAX4250AAUK	68			uB	

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ELECTRICAL CHARACTERISTICS (continued)

 $(V_{DD} = 5V, V_{SS} = 0V, V_{CM} = 0V, V_{OUT} = V_{DD}/2, R_L \text{ connected to } V_{DD}/2, \overline{SHDN} = V_{DD}, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted.}$ Typical values are at $T_A = +25$ °C.) (Notes 2, 3)

PARAMETER	SYMBOL	CONDITIONS			MIN	TYP	MAX	UNITS	
Power-Supply Rejection Ratio	PSRR	V _{DD} – 2.4V to 5.5V	E temperati	ure	75	100		dB	
rower-supply nejection natio	ronn	VDD - 2.4V to 5.5V	MAX4250A	AUK	72			αь	
		$R_L = 10k\Omega$ to $V_{DD}/2$; $V_{OUT} = 25mV$ to V_{DD}	E temperati	E temperature		116			
Laure Ciara I Valta da Oaia		- 4.97V	MAX4250A	AUK	77			-ID	
Large-Signal Voltage Gain	Ay	$R_L = 1k\Omega$ to $V_{DD}/2$;	E temperati	ure	80	112		dB	
		$V_{OUT} = 150V \text{ to } V_{DD}$ $-4.75V$	MAX4250A	AUK	77				
			V _{DD} - V _{OH}	Е		8	25		
Output Voltage Swing	Vout	IV_{IN+} - $V_{IN-}I \ge 10mV$;	VDD VON	Α			30	mV	
- Carpat Foliage Ching	V001	$R_L = 10k\Omega$ to $V_{DD}/2$	V _{OL} - V _{SS}	E A		7	20	1111	
							25		
		$\begin{aligned} & V_{IN+}-V_{IN-}I \geq 10 mV, \\ &R_L = 1 k\Omega \text{ to } V_{DD}/2 \end{aligned}$	V _{DD} - V _{OH}	Е		77	200	mV	
Output Voltage Swing	Vout			A			225		
				E A		47	100 125		
Output Short-Circuit Current	Isc	I A				68	123	mA	
Output Leakage Current	ILEAK	Shutdown mode (SHDN = Vss), Vout = Vss to Vpp (Note 2)				0.001	1.0	μΑ	
SHDN Logic Low	VIL	(Note 2)					0.2 X V _{DD}	V	
SHDN Logic High	VIH	(Note 2)			0.8 X V _{DD}			V	
SHDN Input Current	IIL/IIH	SHDN = V _{SS} = V _{DD} (N	lote 2)			0.5	1.5	μΑ	
Input Capacitance						11		pF	
Gain-Bandwidth Product	GBW	MAX4250-MAX4254				3		MHz	
dain-bandwidti i Toduct	GBW	MAX4249/MAX4255/MAX4256/MAX4257				22		IVI□∠	
Slew Rate	SR	MAX4250-MAX4254				0.3		V/µs	
Olew Hate	On	MAX4249/MAX4255/MAX4256/MAX4257				2.1		ν/μ3	
Peak-to-Peak Input-Noise Voltage	e _{np-} p	f = 0.1Hz to 10Hz				760		nV _{P-P}	
		f = 10Hz				27			
Input Voltage-Noise Density	en	f = 1kHz				8.9		nV/√Hz	
		f = 30kHz				7.9			
Input Current-Noise Density	in	f = 1kHz				0.5		fA/√Hz	

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ELECTRICAL CHARACTERISTICS (continued)

 $(V_{DD} = 5V, V_{SS} = 0V, V_{CM} = 0V, V_{OUT} = V_{DD}/2, R_L \text{ connected to } V_{DD}/2, \overline{SHDN} = V_{DD}, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted.}$ Typical values are at $T_A = +25^{\circ}C$.) (Notes 2, 3)

PARAMETER	SYMBOL	со	NDITIONS	5	MIN	TYP	MAX	UNITS	
		MAX4250-MAX42 A _V = 1V/V, V _{OUT}		f = 1kHz		0.0004			
Total Harmonic Distortion Plus	THD+N	$R_L = 1k\Omega$ to GND (Note 7)		f = 20kHz		0.006		%	
Noise	THEFIN	MAX4249/MAX42 MAX4256/MAX42	=	f = 1kHz		0.0012		/0	
		$A_V = 1V/V, V_{OUT} = 1k\Omega$ to GND		f = 20kHz		0.007			
Capacitive-Load Stability		No sustained osc	illations			400		рF	
		MAX4250-MAX42	254, A _V =	1V/V		10			
Gain Margin	GM	MAX4249/MAX4255/MAX4256/MAX4257, A _V = 10V/V			12.5		dB		
		MAX4250-MAX4254, A _V = 1V/V				74			
Phase Margin	ΦМ	MAX4249/MAX42 A _V = 10V/V	MAX4249/MAX4255/MAX4256/MAX4257, $A_V = 10V/V$		68		Degrees		
		T- 0.010/ Volum	MAX425	0-MAX4254		6.7			
Settling Time		To 0.01%, VOUT = 2V step		9/MAX4255/ 6/MAX4257		1.6		μs	
		IVDD = 5% of	MAX4251/MAX4253			0.8			
Delay Time to Shutdown	tsh	normal operation	NANY 40 40 /NANY 40 FC		1.2			μs	
Delay Time to Enable			MAX4251/MAX4253			8		μs	
Boldy Time to Endois				9/MAX4256		3.5		F -	
Power-Up Delay Time	tpU	V _{DD} = 0 to 5V ste	p, Vout st	able to 0.1%		6		μs	

Note 2: SHDN is available on the MAX4249/MAX4251/MAX4253/MAX4256 only.

Note 3: All device specifications are 100% tested at $T_A = +25$ °C. Limits over temperature are guaranteed by design.

Note 4: Guaranteed by the PSRR test.

Note 5: Offset voltage prior to reflow on the UCSP.

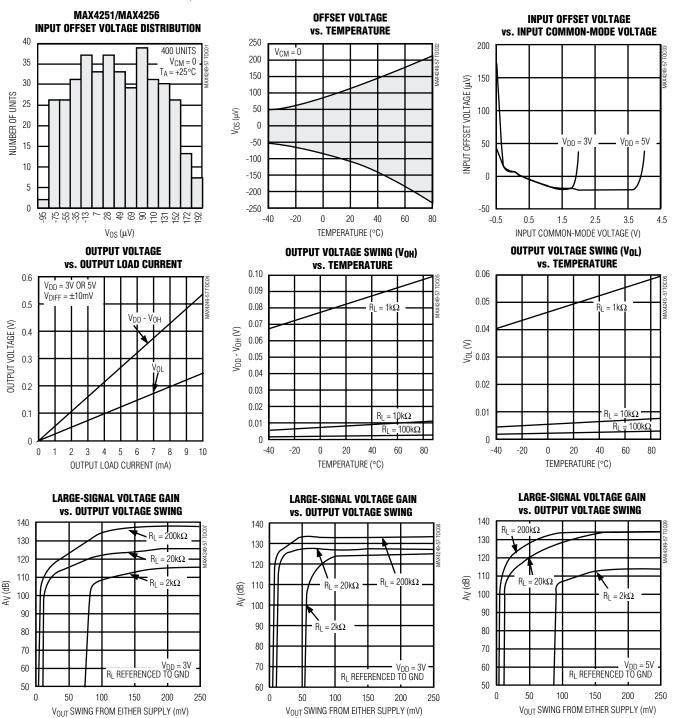
Note 6: Guaranteed by design.

Note 7: Lowpass-filter bandwidth is 22kHz for f = 1kHz and 80kHz for f = 20kHz. Noise floor of test equipment = 10nV/√Hz.

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Typical Operating Characteristics

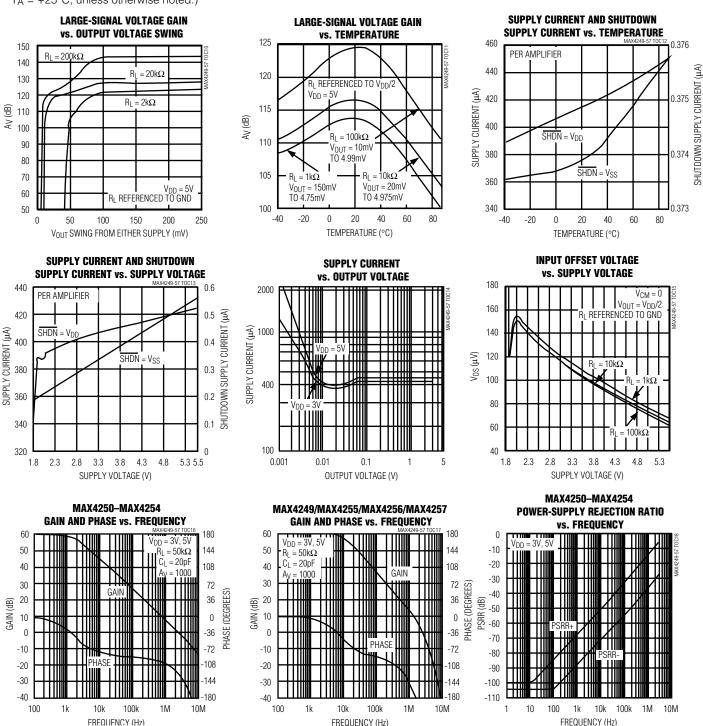
 $(V_{DD} = 5V, V_{SS} = 0V, V_{CM} = V_{DUT} = V_{DD}/2$, input noise floor of test equipment =10nV/ \sqrt{Hz} for all distortion measurements, $T_A = +25^{\circ}C$, unless otherwise noted.)



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Typical Operating Characteristics (continued)

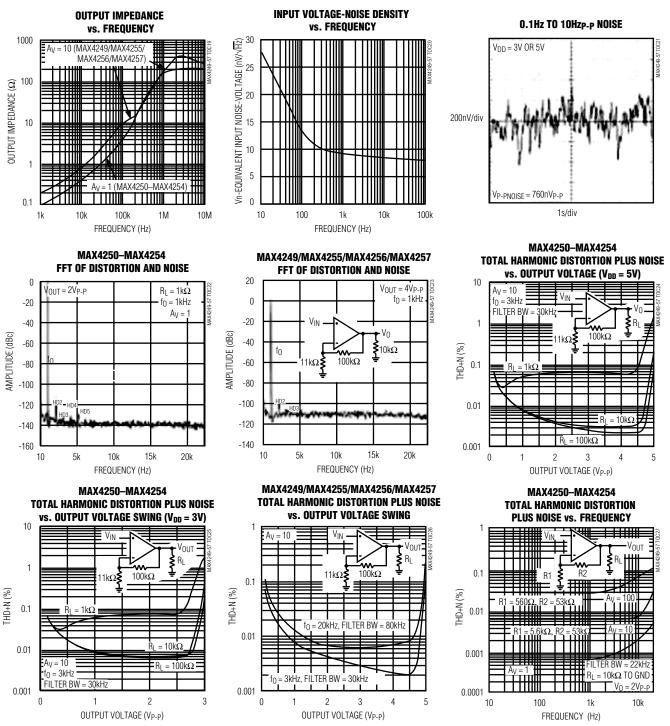
 $(V_{DD} = 5V, V_{SS} = 0V, V_{CM} = V_{DD}/2, input noise floor of test equipment = 10nV/\sqrt{Hz} for all distortion measurements, T_A = +25°C, unless otherwise noted.)$



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Typical Operating Characteristics (continued)

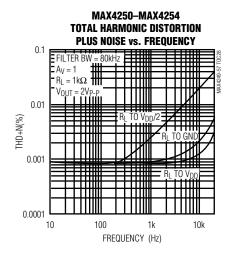
 $(V_{DD} = 5V, V_{SS} = 0V, V_{CM} = V_{DD}/2$, input noise floor of test equipment =10nV/ \sqrt{Hz} for all distortion measurements, $T_A = +25^{\circ}C$, unless otherwise noted.)

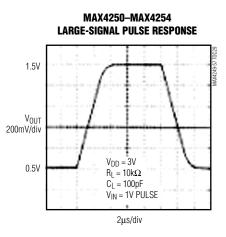


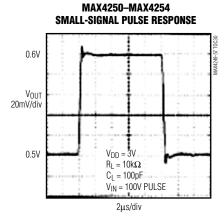
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Typical Operating Characteristics (continued)

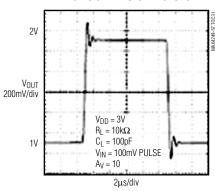
 $(V_{DD} = 5V, V_{SS} = 0V, V_{CM} = V_{DD}/2$, input noise floor of test equipment =10nV/ \sqrt{Hz} for all distortion measurements, $T_A = +25^{\circ}C$, unless otherwise noted.)



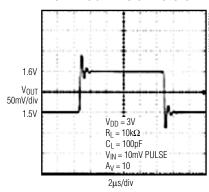


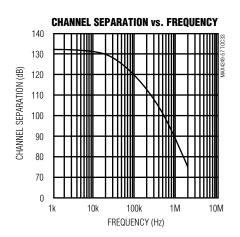


MAX4249/MAX4255/MAX4256/MAX4257 Large-Signal Pulse Response



MAX4249/MAX4255/MAX4256/MAX4257 SMALL-SIGNAL PULSE RESPONSE





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Pin/Bump Description

			PIN/BUM	Р					
MAX4250/ MAX4255	MAX4251/ MAX4256	MAX4252/ MAX4257	MAX4252		MAX4249/ MAX4253		MAX4254	NAME	FUNCTION
5-PIN SOT23	8-PIN SO/μMAX	8-PIN SO/μMAX	8-BUMP UCSP	10-BUMP UCSP	10-PIN μMAX	14-PIN SO	14-PIN SO		
1	6	1, 7	A1, A3	A1, C1	1, 9	1, 13	1, 7, 8, 14	OUT, OUTA, OUTB, OUTC, OUTD	Amplifier Output
2	4	4	C2	B4	4	4	11	Vss	Negative Supply. Connect to ground for single- supply operation
3	3	3, 5	C1, C3	A3, C3	3, 7	3, 11	3, 5, 10, 12	IN+, INA+, INB+, INC+, IND+	Noninverting Amplifier Input
4	2	2, 6	B1, B3	A2, C2	2, 8	2, 12	2, 6, 9, 13	IN-, INA-, INB-, INC-, IND-	Inverting Amplifier Input
5	7	8	A2	B1	10	14	4	V_{DD}	Positive Supply
	8	-	_	A4, C4	5, 6	6, 9	_	SHDN, SHDNA, SHDNB	Shutdown Input, Connect to V _{DD} or leave unconnected for normal operation (amplifier(s) enabled).
_	1, 5	_	_	_	_	5, 7, 8, 10	_	N.C.	No Connection. Not internally connected.
_	_	_	B2	B2, B3	_	_	_	_	Not populated with solder sphere

Detailed Description

The MAX4249–MAX4257 single-supply operational amplifiers feature ultra-low noise and distortion while consuming very little power. Their low distortion and low noise make them ideal for use as preamplifiers in wide dynamic-range applications, such as 16-bit analog-to-digital converters (see *Typical Operating Circuit*). Their high-input impedance and low noise are also useful for signal conditioning of high-impedance sources, such as piezoelectric transducers.

These devices have true rail-to-rail output operation, drive loads as low as $1k\Omega$ while maintaining DC accura-

cy, and can drive capacitive loads up to 400pF without oscillation. The input common-mode voltage range extends from V_{DD} - 1.1V to 200mV beyond the negative rail. The push-pull output stage maintains excellent DC characteristics, while delivering up to ± 5 mA of current.

The MAX4250–4254 are unity-gain stable, whereas, the MAX4249/MAX4255/MAX4256/MAX4257 have a higher slew rate and are stable for gains ≥ 10V/V. The MAX4249/MAX4251/MAX4253/MAX4256 feature a low-power shutdown mode, which reduces the supply current to 0.5µA and disables the outputs.

The MAX4250AAUK is specified for operation over the automotive (-40°C to +125°C) temperature range.

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Low Distortion

Many factors can affect the noise and distortion that the device contributes to the input signal. The following guidelines offer valuable information on the impact of design choices on Total Harmonic Distortion (THD).

Choosing proper feedback and gain resistor values for a particular application can be a very important factor in reducing THD. In general, the smaller the closed-loop gain, the smaller the THD generated, especially when driving heavy resistive loads. Large-value feedback resistors can significantly improve distortion. The THD of the part normally increases at approximately 20dB per decade, as a function of frequency. Operating the device near or above the full-power bandwidth significantly degrades distortion.

Referencing the load to either supply also improves the part's distortion performance, because only one of the MOSFETs of the push-pull output stage drives the output. Referencing the load to midsupply increases the part's distortion for a given load and feedback setting. (See the Total Harmonic Distortion vs. Frequency graph in the *Typical Operating Characteristics*.)

For gains ≥ 10V/V, the decompensated devices MAX4249/MAX4255/MAX4256/MAX4257 deliver the best distortion performance, since they have a higher slew rate and provide a higher amount of loop gain for a given closed-loop gain setting. Capacitive loads below 400pF, do not significantly affect distortion results. Distortion performance remains relatively constant over supply voltages.

Low Noise

The amplifier's input-referred, noise-voltage density is dominated by flicker noise at lower frequencies, and by thermal noise at higher frequencies. Because the thermal noise contribution is affected by the parallel combination of the feedback resistive network (RF II RG, Figure 1), these resistors should be reduced in cases where the system bandwidth is large and thermal noise is dominant. This noise contribution factor decreases, however, with increasing gain settings.

For example, the input noise-voltage density of the circuit with RF = $100k\Omega$, RG = $11k\Omega$ (AV = 10V/V) is en = 15nV/VHz, en can be reduced to 9nV/VHz by choosing RF = $10k\Omega$, RG = $1.1k\Omega$ (AV = 10V/V), at the expense of greater current consumption and potentially higher distortion. For a gain of 100V/V with RF = $100k\Omega$, RG = $1.1k\Omega$, the en is low (9nV/VHz).

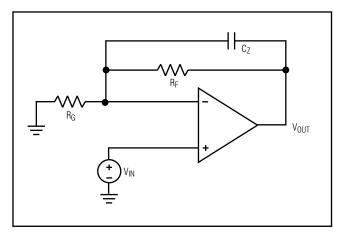


Figure 1. Adding Feed-Forward Compensation

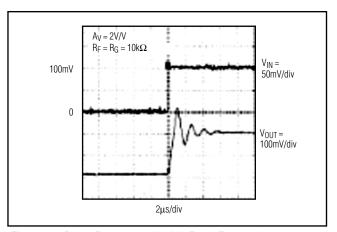


Figure 2a. Pulse Response with No Feed-Forward Compensation

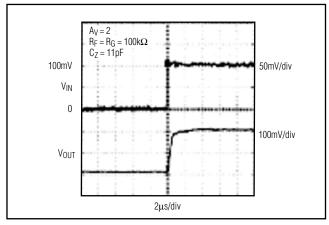


Figure 2b. Pulse Response with 10pF Feed-Forward Compensation

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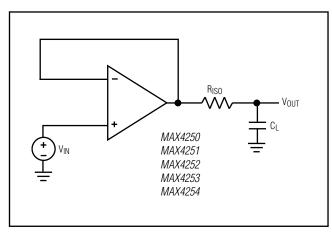


Figure 3. Overdriven Input Showing No Phase Reversal

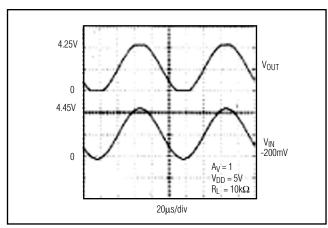


Figure 4. Rail-to-Rail Output Operation

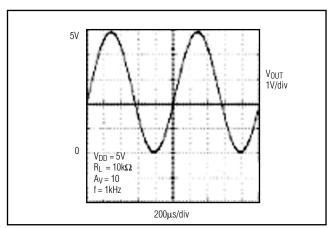


Figure 5. Capacitive-Load Driving Circuit

Using a Feed-Forward Compensation Capacitor, Cz

The amplifier's input capacitance is 11pF. If the resistance seen by the inverting input is large (feedback network), this can introduce a pole within the amplifier's bandwidth, resulting in reduced phase margin. Compensate the reduced phase margin by introducing a feed-forward capacitor (Cz) between the inverting input and the output (Figure 1). This effectively cancels the pole from the inverting input of the amplifier. Choose the value of Cz as follows:

$$C_Z = 11 \times (R_F / R_G) [pF]$$

In the unity-gain stable MAX4250–MAX4254, the use of a proper Cz is most important for Av = 2V/V, and Av = -1V/V. In the decompensated MAX4249/MAX4255/MAX4256/MAX4257, Cz is most important for Av = 10V/V. Figures 2a and 2b show transient response both with and without Cz.

Using a slightly smaller Cz than suggested by the formula above achieves a higher bandwidth at the expense of reduced phase and gain margin. As a general guideline, consider using Cz for cases where Rg II RF is greater than $20k\Omega$ (MAX4250–MAX4254) or greater than $5k\Omega$ (MAX4249/MAX4255/MAX4256/MAX4257).

Applications Information

The MAX4249–MAX4257 combine good driving capability with ground-sensing input and rail-to-rail output operation. With their low distortion, low noise, and low-power consumption, these devices are ideal for use in portable instrumentation systems and other low-power, noise-sensitive applications.

Ground-Sensing and Rail-to-Rail Outputs

The common-mode input range of these devices extends below ground, and offers excellent common-mode rejection. These devices are guaranteed not to undergo phase reversal when the input is overdriven (Figure 3).

Figure 4 showcases the true rail-to-rail output operation of the amplifier, configured with Av = 10V/V. The output swings to within 8mV of the supplies with a $10k\Omega$ load, making the devices ideal in low-supply-voltage applications.

Output Loading and Stability

Even with their low quiescent current of 400 μ A, these amplifiers can drive 1k Ω loads while maintaining excellent DC accuracy. Stability while driving heavy capacitive loads is another key feature.

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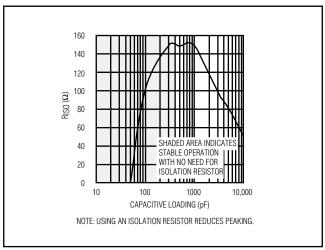


Figure 6. Isolation Resistance vs. Capacitive Loading to Minimize Peaking (<2dB)

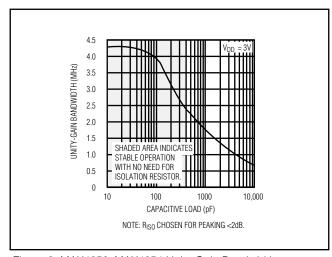


Figure 8. MAX4250–MAX4254 Unity-Gain Bandwidth vs. Capacitive Load

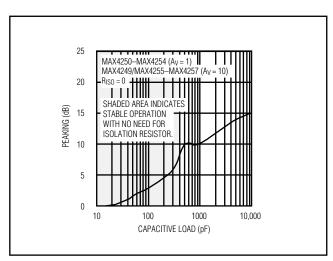


Figure 7. Peaking vs. Capacitive Load

These devices maintain stability while driving loads up to 400pF. To drive higher capacitive loads, place a small isolation resistor in series between the output of the amplifier and the capacitive load (Figure 5). This resistor improves the amplifier's phase margin by isolating the capacitor from the op amp's output. Reference Figure 6 to select a resistance value that will ensure a load capacitance that limits peaking to <2dB (25%). For example, if the capacitive load is 1000pF, the corresponding isolation resistor is 150 Ω . Figure 7 shows that peaking occurs without the isolation resistor. Figure 8 shows the unity-gain bandwidth vs. capacitive load for the MAX4250–MAX4254.

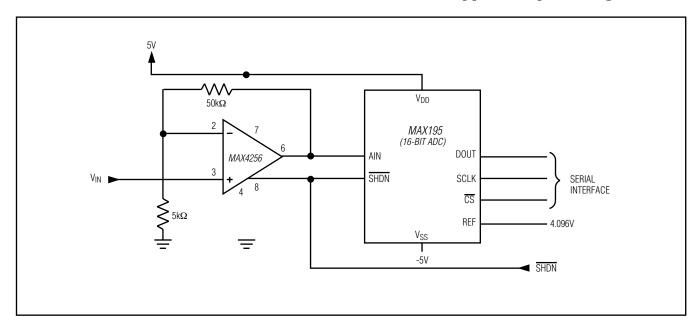
Power Supplies and Layout

The MAX4249–MAX4257 operate from a single 2.4V to 5.5V power supply or from dual supplies of ± 1.20 V to ± 2.75 V. For single-supply operation, bypass the power supply with a 0.1µF ceramic capacitor placed close to the V_{DD} pin. If operating from dual supplies, bypass each supply to ground.

Good layout improves performance by decreasing the amount of stray capacitance and noise at the op amp's inputs and output. To decrease stray capacitance, minimize PC board trace lengths and resistor leads, and place external components close to the op amp's pins.

UCSP, Single-Supply, Low-Noise, Low-Distortion, Rail-to-Rail Op Amps

Typical Operating Circuit

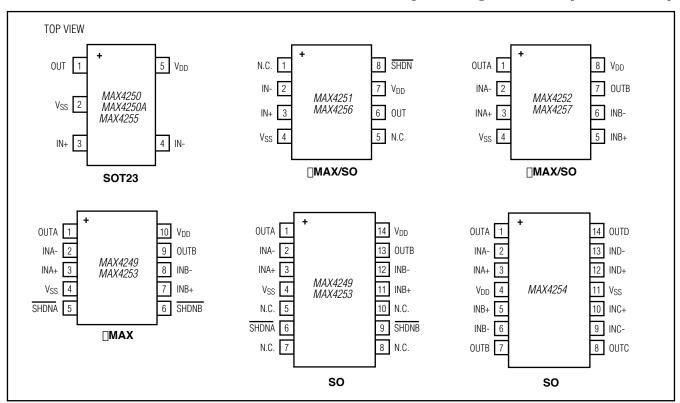


Selector Guide

PART	GAIN BANDWIDTH (MHz)	MINIMUM STABLE GAIN (V/V)	NO. OF AMPLIFIERS PER PACKAGE	SHUTDOWN MODE	PIN-PACKAGE
MAX4249	22	10	2	Yes	10-pin μMAX, 14-pin SO
MAX4250/A	3	1	1		5-pin SOT23
MAX4251	3	1	1	Yes	8-pin μMAX/SO
MAX4252	3	1	2	_	8-pin µMAX/SO, 8-bump UCSP
MAX4253	3	1	2	Yes	10-pin µMAX, 14-pin SO, 10-bump UCSP
MAX4254	3	1	4	_	14-pin SO
MAX4255	22	10	1	_	5-pin SOT23
MAX4256	22	10	1	Yes	8-pin μMAX/SO
MAX4257	22	10	2	_	8-pin μMAX/SO

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Pin/Bump Configurations (continued)



_Ordering Information (continued)

PART	TEMP RANGE	PIN- PACKAGE	TOP MARK
MAX4251ESA+	-40°C to +85°C	8 SO	_
MAX4251EUA+	-40°C to +85°C	8 µMAX	_
MAX4252EBL+T	-40°C to +85°C	8 UCSP	AAO
MAX4252ESA+	-40°C to +85°C	8 SO	_
MAX4252EUA+	-40°C to +85°C	8 µMAX	_
MAX4253EBC+T	-40°C to +85°C	10 UCSP	AAK
MAX4253EUB+	-40°C to +85°C	10 μMAX	_
MAX4253ESD+	-40°C to +85°C	14 SO	_
MAX4254ESD+	-40°C to +85°C	14 SO	_
MAX4255EUK+T	-40°C to +85°C	5 SOT23	ACCJ
MAX4256ESA+	-40°C to +85°C	8 SO	_
MAX4256EUA+	-40°C to +85°C	8 µMAX	_
MAX4257ESA+	-40°C to +85°C	8 SO	_
MAX4257ESA/V+T	-40°C to +85°C	8 SO	_
MAX4257EUA+	-40°C to +85°C	8 µMAX	

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Package Information

For the latest package outline information and land patterns (footprints), go to www.maximintegrated.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	OUTLINE NO.	LAND PATTERN NO.
5 SOT-23	U5+2	<u>21-0057</u>	<u>90-0174</u>
8 μMAX	U8+1	<u>21-0036</u>	<u>90-0092</u>
10 μMAX	U10+2	<u>21-0061</u>	<u>90-0330</u>
3 x 3 μCSP	B9+5	<u>21-0093</u>	_
14 SOIC	S14+1	<u>21-0041</u>	<u>90-0112</u>
12 µCSP	B12+4	<u>21-0104</u>	_

UCSP, Single-Supply, Low-Noise, Low-Distortion, Rail-to-Rail Op Amps

Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
8	10/11	Added lead-free packaging to the <i>Ordering Information</i> and changed the Input Bias Current and Input Offset Current conditions in the <i>Electrical Characteristics</i> table	1, 2, 14
9	12/12	Added MAX4257ESA/V+T to Ordering Information.	14



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